

## EXPERIMENT #3 TRANSISTOR BIASING

Bias (operating point) for a transistor is established by specifying the quiescent (D.C., no signal) values of collector-emitter voltage  $V_{CEQ}$  and collector current  $I_{CQ}$ . Reliable operation of a transistor over a wide range of temperatures requires that bias voltage and current remain stable. However, variations of reverse-bias collector current  $I_{CO}$ , and emitter-base junction voltage with temperature preclude stable bias unless external compensating circuits are used. Bias stabilizing circuits may employ resistors, thermistors, diodes, etc.

The choice of the operating point of a transistor is determined by several factors, such as maximum voltage swing, allowable operating region and small signal parameters. If the  $Q$ -point shifts, the output signal might get clipped or the transistor may go out of the safe operating region. *Figure 2-1* shows three types of transistor biasing circuits. The configuration shown in the *Figure 2-1a* is the simplest way of biasing a bipolar transistor. There is no stabilization in this circuit hence any change in the transistor parameters or the ambient temperature will shift the  $Q$ -point. For a desired  $V_{CEQ}$  and  $I_{CQ}$  pair, the resistor values may be calculated as follows:

$$R_B = h_{FE} \frac{V_{CC} - 0.6}{I_{CQ}} \quad (2.1)$$

$$R_C = \frac{V_{CC} - V_{CEQ}}{I_{CQ}} \quad (2.2)$$

In order to stabilize the operating point, some kind of negative feedback must be used at the expense of a reduced voltage gain. *Figure 2-1b* shows one of the simplest bias circuits of this kind. If the  $Q$ -point of the circuit tends to shift, the base current, which is proportional to  $V_{CE}$ , will increase or decrease accordingly to compensate for this shift. The equation (2.2) is also valid for this circuit provided that  $h_{FE} \gg 1$  (which is true for most modern transistors).  $R_B$  can then be obtained from the equation (2.3).

$$R_B = h_{FE} \frac{V_{CEQ} - 0.6}{I_{CQ}} \quad (2.3)$$

The inclusion of an emitter resistance always improves the bias stabilization. But the collector current and the voltage are still depending on  $h_{FE}$ . In order to get a bias, which is independent of transistor parameters, the base of the transistor must be driven from a voltage source. That is, the equivalent output impedance of the source must be low compared to the input impedance of the transistor. If,

$$R_B = \frac{R_{B1}R_{B2}}{R_{B1} + R_{B2}} \ll h_{FE} R_E \quad (2.4)$$

this condition is satisfied in the circuit in the *Figure 2-1c*. The voltage source to drive the base is obtained using voltage dividing resistors  $R_{B1}$ ,  $R_{B2}$ . Assuming the bleeder current,  $I_1$ , is ten times larger than the base current,  $I_B = I_C / h_{FE}$ , (this assumption usually satisfies the equation 2.4) the resistor values may be easily evaluated as follows:

$$R_E = \frac{V_E}{I_{CQ}} \quad (2.5)$$

$$R_C = \frac{V_{CC} - V_{CEQ} - V_E}{I_{CQ}} \quad (2.6)$$

$$R_{B1} = h_{FE} \frac{V_{CC} - V_E - 0.6}{10I_{CQ}} \quad (2.7)$$

$$R_{B2} = h_{FE} \frac{V_E + 0.6}{10I_{CQ}} \quad (2.8)$$

The quiescent voltages and currents of this circuit are quite independent of transistor parameters and temperature and are functions of resistance values only.

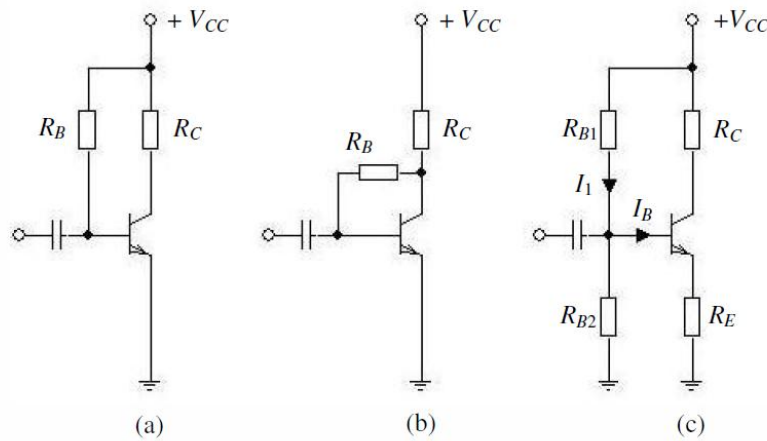


Figure 2-1: Basic Bipolar Transistor biasing circuits

## EQUIPMENT

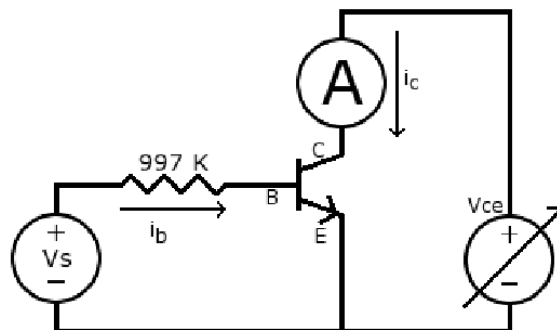
1. CRT oscilloscope
2. Electronic D.C. voltmeter
3. D.C. power supply
4. Sine-wave generator
5. Breadboard

## COMPONENTS

1. BC238B or equivalent transistors
2. Resistors (to be calculated)
3. Capacitor ( $0.1\mu\text{F}$ )

## PROCEDURE

The first step in creating the data charts for measuring the BJT's characteristics is to construct a test circuit. Here, we are using a simple circuit with only one resistor, two voltage supplies, an ammeter, and the BJT. The circuit was constructed with the following schematic:



We will use three experimental trials, with a different  $V_s$  for each trial. Since the voltage drop across the Base-Emitter junction is a constant 0.7 volts, the remainder of  $V_s$  must drop across the resistor. This allows us to find the value of the base current. The resistor used was labeled  $1\text{ M}\Omega$ , but the real value was  $997\text{ K}\Omega$ . For each of the three trials ( $V_{BE}=20\text{V}$ ,  $25\text{V}$  and  $30\text{V}$ ), vary  $V_{CE}$  between 0.02 volts and 20 volts and measure the resulting collector current,  $I_C$ . Fill the following table with the data you collected.

**Table 2.1**

	COLLECTOR CURRENT ( $I_C$ )		
$V_{CE}$ (Volts)	$V_S=20$ V	$V_S=25$ V	$V_S=30$ V
0.02			
0.04			
0.06			
0.08			
0.1			
0.12			
0.14			
0.16			
0.18			
0.2			
0.3			
0.4			
0.5			
1.0			
3.0			
5.0			
7.0			
9.0			
11.0			
13.0			
15.0			
17.0			
19.0			
20.0			

After finishing the first step continue with the following steps:

1. a) Calculate the values of  $R_B$  and  $R_C$  in *Figure 2-1a* for  $V_{CC} = 12V$ ,  $V_{CEQ} = 6V$ ;  $I_{CQ} = 1mA$  and  $h_{FE} = 200$ .

$$R_B = \dots\dots\dots$$

$$R_C = \dots\dots\dots$$

- b) Setup the circuit. Use the nearest standard values for resistors. Measure and note the collector current  $I_{CQ}$  and collector emitter voltage  $V_{CEQ}$ . Calculate the real  $h_{FE}$  of the transistor, using the measured  $I_{CQ}$ .

$$I_{CQ} = \dots\dots\dots$$

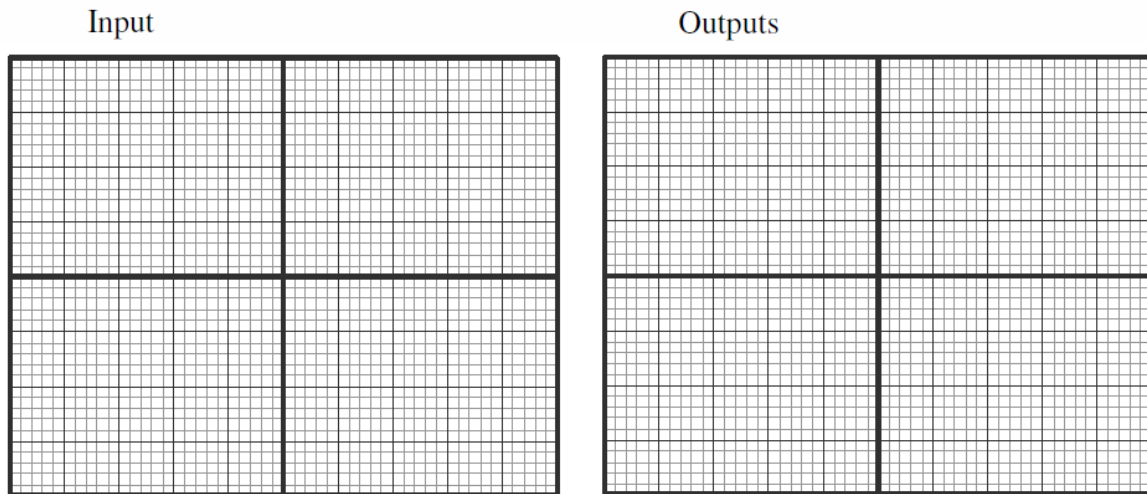
$$V_{CEQ} = \dots\dots\dots$$

$$h_{FE} = \dots\dots\dots$$

- c) Exchange your transistor with other specimens of the same type. Measure once more  $V_{CEQ}$  and  $I_{CQ}$  for at least four different transistors. Fill in the blanks on *Table-2.1*.

- d) Connect an oscilloscope to the collector of the transistor and apply a 10kHz sinusoidal voltage to the base through a 0.1μF capacitor, increase the input voltage until any perceptible distortion occurs at the output signal. Then reduce the signal to one half and measure the peak-to-peak input and output voltages on the oscilloscope. Calculate the voltage gain. Increase the input voltage again until a small distortion occurs at the output signal. Then change your transistor and sketch the waveforms.

$$A_V(a) = \dots\dots\dots$$



2. a) Calculate  $R_B$  and  $R_C$  of *Figure 2-1b* to obtain the same operating points as in *1-a*.

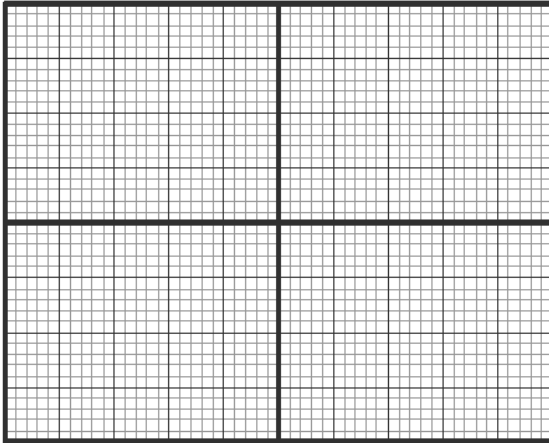
$$R_B = \dots\dots\dots$$

$$R_C = \dots\dots\dots$$

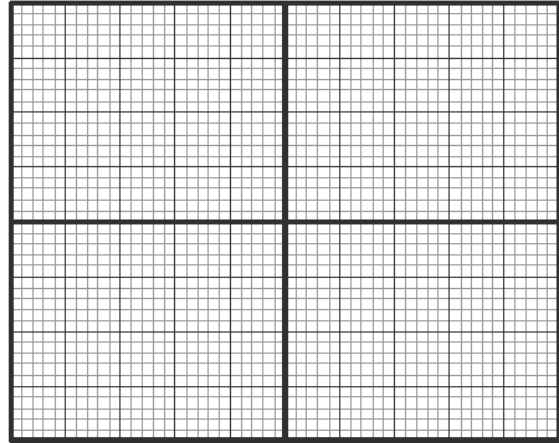
- b) Repeat *1-b*, *1-c* and *1-d*. Fill in the blanks in *Table-2.1*.

$$A_V(b) = \dots\dots\dots$$

Input



Outputs



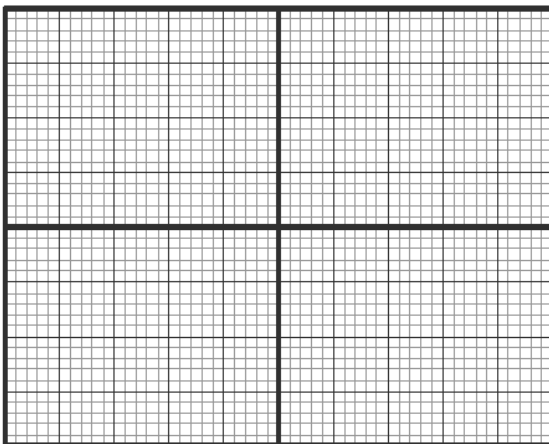
3. a) Calculate all resistor values of *Figure 2-1c*, for  $V_{CEQ} = 5.5V$ ;  $I_{CQ} = 1mA$ ;  $V_E = 1V$

$R_{B1} = \dots\dots\dots$        $R_{B2} = \dots\dots\dots$        $R_C = \dots\dots\dots$        $R_E = \dots\dots\dots$

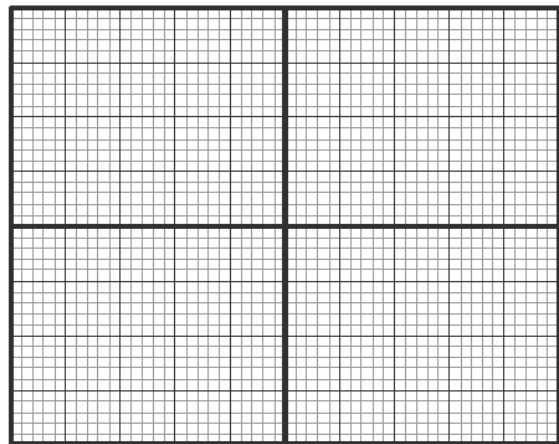
- b) Repeat 1-b, 1-c and 1-d. Fill in the blanks in *Table-2.1*.

$A_V(c) = \dots\dots\dots$

Input



Outputs



**Table 2-2**

Figure #	2-a		2-b		2-c	
Specimen #	$V_{CEQ}$	$I_{CQ}$	$V_{CEQ}$	$I_{CQ}$	$V_{CEQ}$	$I_{CQ}$
1						
2						
3						
4						
Max. Variation $ X_{max}-X_{min} $						

Note: Xmax in Table 2.1 is the maximum variation of the 4 specimens. For example  $V_{CEQ}$  is maximum for specimen 2 and it is minimum for specimen 4. Then Max. Variation will be  $V_{CEQ(2)} - V_{CEQ(4)}$ .